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RADAR & sensor basics

HBAR

Increasing operating frequency

Conclusion

High-overtone Bulk Acoustic Resonator (HBAR) as passive sensor: towards microwave wireless interrogation

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Introduction

Context: acoustic wave transducers used as passive, wireless sensors

Objective: use of commercially available RADAR for probing cooperative target acting as sensor

- Alternative strategy to energy harvesting: no energy at all !
- Passive transducer acts as sensor remotely characterized
- The sensor itself is tiny ($<5 \times 5 \text{ mm}^2$) but the antenna is huge
- Analog transducer does not provide identification or anticollision capability

 \Rightarrow increase operating range to the microwave range for **reduced antenna size** and **spatial multiplexing** thanks to directive beams with modest antenna dimensions on the reader.

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Basics of RADAR

- bistatic (physically separated emitter and receiver) or monostatic configurations: isolation defines rage
- echos due to electromagnetic impedance variations (permittivity ε_r and conductivity σ)

$$v = \frac{c}{\sqrt{\frac{\varepsilon_r}{2} \left(\sqrt{1 + \frac{\sigma^2}{\varepsilon^2 \omega^2}} + 1\right)}}$$

- provides both magnitude *and phase* informations on the returned pulse
- typical frequency range: 50 MHz-50 GHz ^a

 $^a{\rm H.}$ Stockman, Communication by means of reflected power, Proc. I.R.E ${\bf 36}$ pp.1196-1204 (1948)

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RADAR example (2)

(a)

HF-VHF RADAR is long range (even over the horizon), but requires excessive antenna dimensions for industrial applications $(\lambda_m = 300/f_{MHz} \Rightarrow \lambda/4 = 1 \text{ m at 75 MHz}).$



Objective: electromagnetic scanvengers, here called cooperative target

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Basics of Surface Acoustic Wave (SAW) delay lines

- acoustic = propagation of a mechanical wave on a substrate
- most efficient way of converting electromagnetic (EM) to mechanical: piezoelectric substrate + interdigitated transducers
- identification + sensor
- physical quantity measurement function of acoustic velocity
- incoming EM pulse generates mechanical pulse which returns as EM with a time delay function of physical quantity (temperature, stress, pressure ...)



- high electromechanical coupling coefficient (LNO)
- mirror = patterned electrodes
- time delay between incoming pulse and reflection = measurement
- typical velocity: 1500-5000 m/s for most materials
- typical delays: 1-5 μ s (3 μ s at 3000 m/s \Rightarrow 4.5 mm path)

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SAW delay line as RADAR cooperative target

Acoustic transducer as RADAR cooperative target:

- complement the passive interface monitoring with sensor interrogation
- linear conversion process from EM to mechanical: no threshold voltage (cf diodes in Si based RFID)



Challenge: at 5000 m/s, a sensor operating at 5 GHz would require 250 nm lithography (with $\ll \lambda$ resolution)

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Link budget for delay lines

- RADAR illumination of point-like target: decay as $1/d^4$
- Free Space Propagation Loss (FSPL)

$$10 \times \log_{10} \left(\frac{\lambda^2}{4\pi} \times \frac{\lambda^2}{4\pi} \times \frac{1}{\left(4\pi d^2\right)^2} \right) = 10 \log_{10} \left(\frac{\lambda^4}{(4\pi)^4 d^4} \right)$$

• Considering we know the range at ice-rock interface and reflection coeffient

$$\left(rac{arepsilon_{\it ice} - arepsilon_{\it rock}}{arepsilon_{\it ice} + arepsilon_{\it rock}}
ight)^2 \simeq 19 \; {
m dB}$$

$$\Rightarrow d_{SAW} = d_{ice-rock} imes 10^{(IL_{ice-rock} - IL_{SAW})/40} \simeq$$
 40 m

assuming $d_{ice-rock} = 100$ m, consistent with SNR of a 5 m deep-measurement 1

¹J.-M Friedt, T. Rétornaz, S. Alzuaga, T. Baron, G. Martin, T. Laroche, S. Ballandras, M. Griselin & J.-P. Simonnet, *Surface Acoustic Wave Devices as Passive Buried Sensors* J. Appl. Phys. **109** (3), pp. 034905 (2011) ((B) + (E) +

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High-overtone Bulk Acoustic Resonator (HBAR)

 The acoustic wave no longer propagates at the air-crystal interface but in the bulk of the crystal

- Operating frequencies are defined by layer thicknesses rather than lithography of electrodes
- Oly-crystalline active layer (AIN, ZnO) or single-crystal (lithium niobate): high coupling
- Low loss propagation substrate exhibiting appropriate sensitivity to the measured quantity



Typical dimensions: 5-10 μm thick piezo, 300-500 μm thick substrate, 2×2 mm^2 chip

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Conclusion

High-overtone Bulk Acoustic Resonator (HBAR)

- The acoustic wave no longer propagates at the air-crystal interface but in the bulk of the crystal
- Operating frequencies are defined by layer thicknesses rather than lithography of electrodes
- Poly-crystalline active layer (AIN, ZnO) or single-crystal (lithium niobate): high coupling
- Low loss propagation substrate Frequency comb from 50 to 500 MHz exhibiting appropriate sensitivity to the measured quantity



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High-overtone Bulk Acoustic Resonator (HBAR)

- The acoustic wave no longer propagates at the air-crystal interface but in the bulk of the crystal
- 2 Operating frequencies are defined by layer thicknesses rather than lithography of electrodes
- 3 Poly-crystalline active layer (AIN, ZnO) or single-crystal (lithium niobate): high coupling

quantity

- time (s) 4 Low loss propagation substrate Time domain echos, 0.5-2.2 μ s exhibiting appropriate sensitivity to the measured
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HBAR measurement strategy

- HBAR spectrum is a comb (in time or frequency domain) of modes
- Frequency domain (resonance identification) or time domain (pulse delay) Digital signal post-processing, *no modification* of RADAR hardware
- Acoustic velocity change with physical property
- no need to change RADAR hardware, only signal post-processing step

Frequency domain caracterisation: incompatible with FMCW RADAR (sweep rate $\ll Q/\pi$ periods) and pulse mode (unable to recover an accurate frequency)

 \Rightarrow time domain approach, search for time delay between returned echos (magnitude & phase)



HBAR



(VHF)

Experimental demonstration (VHF)

Temperature measurement

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Experimental demonstration (VHF)

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Towards microwaves: mixing

Using a VHF transducer at microwave frequencies: use a diode next to the sensor as AM demodulator

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Strategy compatible with **electronic beam steering** on the emission (Space-division multiple access) and omnidirectional receiving antenna (日) (同) (日) (日)

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Towards microwaves: mixing

Using a VHF transducer at microwave frequencies: use a diode next to the sensor as AM demodulator

wireless interrogation J.M Friedt & al.

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Towards microwaves: mixing

interrogation Using a VHF transducer at microwave frequencies: use a diode next to J.M Friedt & al. the sensor as AM demodulator



off resonance

at resonance

Strategy compatible with **electronic beam steering** on the emission (Space-division multiple access) and omnidirectional receiving antenna

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Towards microwaves: baseband

- However, adding a rectifying diode brings back the drawback of RFID
- HBAR can reach the microwave frequency range ... if appropriately designed



In this example, the SU8 assembling glue acts as a strong acoustic reflector and generates modes up to 4 GHz

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- use of a widely available tool (RADAR) for probing sensors ("cooperative targets")
- piezoelectric-based (linear) transducers for improved interrogation range
- signal processing for (time-based) delay line: temperature
- $\bullet \ \Rightarrow$ acoustic delay lines for tagging or sensor applications
- \Rightarrow HBAR for multimode (multiple RADAR instrument) & time-domain interrogation

A **passive** sensor solves the issue of **local** energy harvesting, and moves the energy requirement to the interrogating RADAR \Rightarrow best suited in environments where sensor maintenance is impossible once installed (buried in plastic, concrete, soil ...)





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