

Theoretical & experimental analysis of high Q SAW resonator transient response in a wireless sensor interrogation application

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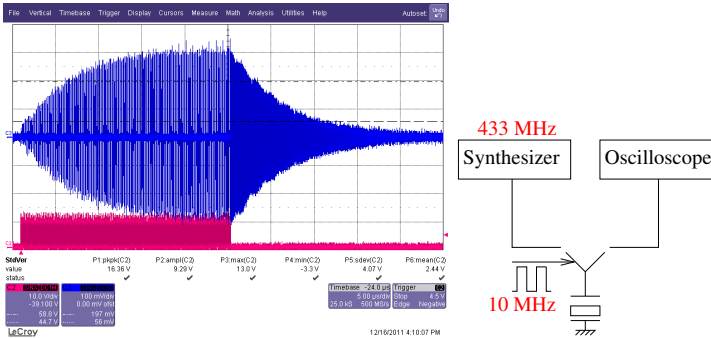


Context: surface acoustic wave (SAW) devices used as **passive sensors** interrogated through a **wireless link** using a **monostatic, frequency-sweep pulsed RADAR**

Objective: **link budget** and **frequency identification resolution** including electronic component noise and acoustic device characteristics in order to identify limiting factor in interrogation range

Oscillator noise limitation

- Phase noise S_φ of local oscillator (LO) = frequency fluctuation of the reference signal: $S_\varphi(f_c) = \frac{\Delta\varphi_{RMS}}{BW}$ (BW: measurement bandwidth, f_c offset from carrier)
- The longer the interrogation, the higher the chance of phase fluctuations (=high phase noise close to carrier)
- resonator time constant $\tau = Q/(\pi f_0) \simeq 7 \mu s$ for resonator at $f_0 = 433 \text{ MHz}$ and quality factor $Q = 10000$: wait for 5τ to reach 99% of load/unload (independent sequence of measurements)
- frequency sweep strategy: $N = 128$ points/frequency band (fixed frequency comb^a) or 2-point strategy (feedback on measured frequency^b)
 $\Rightarrow S_\varphi$ at $f_c = 1/(2 \times N \times \tau) \in [0.1 - 7] \text{ kHz}$



Experimental load/unload of resonator: the carrier is chopped during load so that the stored energy is visible

Phase to frequency noise: $S_{\Delta f} = f_c^2 S_\varphi = \frac{\Delta f_{RMS}^2}{BW}$ i.e. for a measurement bandwidth BW of $2 \times f_c$,

$$\Delta f_{RMS} = \sqrt{2 \times f_c^3 \times S_\varphi}$$

| Phase noise | f_0 | Q | Δf_{RMS} | resolution |
|-------------|-----------|-------|------------------|-----------------------|
| -170 dBc/Hz | 434 MHz | 10000 | 0.01 Hz | 4.10^{-6} K |
| -130 dBc/Hz | 434 MHz | 10000 | 1 Hz | 4.10^{-4} K |
| -90 dBc/Hz | 434 MHz | 10000 | 140 Hz | 50 mK |
| -170 dBc/Hz | 2.450 GHz | 1500 | 2 Hz | 10^{-4} K |
| -130 dBc/Hz | 2.450 GHz | 1500 | 230 Hz | 10 mK |
| -90 dBc/Hz | 2.450 GHz | 1500 | 23000 Hz | 1 K |

assuming 6 ppm/K (200 K range within 434 MHz ISM band). Receiver amplifier noise floor: kT/P (P received power, temperature T) dominates at large range, -114 dBc/Hz if $P = -60 \text{ dBm}$

^aJ.-M Friedt, C. Droit, G. Martin, and S. Ballandras *A wireless interrogation system exploiting narrowband acoustic resonator for remote physical quantity measurement*, Rev. Sci. Instrum. **81**, 014701 (2010)

^bJ.-M Friedt, C. Droit, S. Ballandras, S. Alzuaga, G. Martin, P. Sandoz *Remote vibration measurement: a wireless passive surface acoustic wave resonator fast probing strategy*, Rev. Sci. Instrum. **83**, p.055001 (2012)

SAW transducer limitation

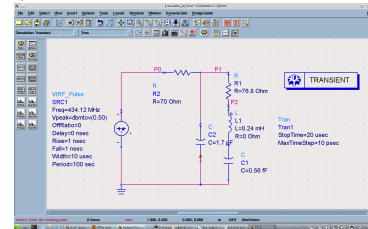
Two approaches:

- finite element modelling followed by mixed matrix assembly of elementary structures (transducer, cavity) $\rightarrow S_{11}$ response file
- equivalent electrical model (Butterworth-van Dyke) with fitted motional parameters

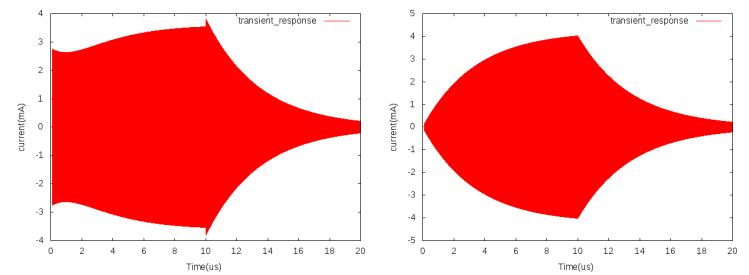
Modelling sequence:

1. include either model in ADS in addition to RADAR electronics component models, antenna impedance set to 70Ω load (to be improved later)
2. time-domain analysis for transient response of acoustic device probed by pulse with variable frequency carrier
3. record amplitude at time $t \ll Q/(\pi f_0)$ for resonator operating at frequency $f_0 \simeq 433 \text{ MHz}$ and quality factor $Q \simeq 10000$

Challenge: high quality factor \Rightarrow simulation time step $\ll 1/f_0$ (here 10 ps steps for a total 10 μs simulation)

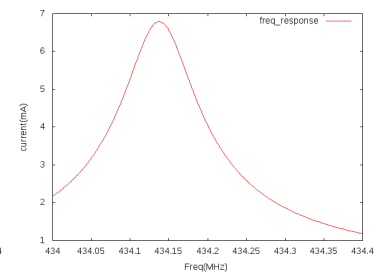
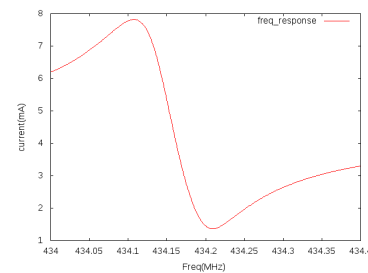


ADS model of the resonator with gated source: source at 434.12 MHz. BVD equivalent circuit: $L_m = 0.24 \text{ mH}$, $C_m = 0.56 \text{ fF}$, $R_m = 76.8 \Omega$, $C_0 = 1.7 \text{ pF}$



Current through BvD (load & unload)

Current through motional branch



Frequency dependent current through BvD (\leftarrow) and motional branch (\rightarrow)
 Current in motional branch = mechanically stored energy (current continuity condition is met)

Conclusion: towards full wireless SAW reader modelling tool, including phase noise of LO and contribution of noise of electronics & transducer. LO is limiting at low range, receiver noise at range $> 1 \text{ m}$.

Perspectives: add propagation path/medium losses (buried sensors) and antenna impedance/radiation pattern.

